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Patent Application

Title: DIRECTIONAL HEARING AID
TESTER

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Title: Directional Hearing Aid Tester

FIELD OF THE INVENTION

[0001] This invention relates to apparatus and methods for testing
5 directional responding acoustical devices to determine their response to
sound stimuli. The directional responding acoustical devices will usually,
although not necessarily, be directional hearing aids.

BACKGROUND OF THE INVENTION

10 **[0002]** Hearing aids are tested by supplying a known acoustical test
stimulus to the hearing aid microphone and measuring the resulting output.
Increasingly, modern hearing aids employ a combination of directional
responding microphones and non-linear signal processing to provide better
performance to the end-user. Because the non-linear circuitry often causes
15 both the gain and the frequency response of the hearing aid to be level-
dependent, it is not possible to measure an accurate directional response by
using (for example) two sound sources, one front-facing (i.e. in front of the
hearing aid) and the other rear-facing (i.e. facing the rear of the hearing aid),
and separately in time sweeping them through various frequencies. An
20 accurate measurement of the directional characteristic requires that the front-
facing and rear-facing acoustical stimuli be presented simultaneously.

[0003] Traditionally, directional response testing for directional hearing
aids has been performed in an anechoic test space in which the front-facing
and rear-facing responses are measured separately, typically by making a
25 front-facing measurement and then rotating the hearing aid 180° in the test
space to make the rear-facing measurement. As mentioned, measuring the
front-facing and rear-facing responses separately will introduce significant
error if the hearing aid has level-dependent gain and frequency shaping
circuitry that responds to the overall input level. As an example, the rear-
30 facing signal may be attenuated by the directional microphone by upwards of
10 dB, so when this signal is presented in isolation, the level-dependent
circuitry will adapt accordingly to this low-level signal. However, under real-

world conditions, the front-facing signal will be present simultaneously with the rear-facing signal and will not be attenuated. This will result in a significantly higher total signal presented to the level-dependent circuitry and consequently the hearing aid will under these conditions have a different gain and frequency response.

[0004] Using an anechoic test space presents additional problems. Such space must be large and filled with sound absorbing material to prevent standing waves, and this makes it impractical for use by most hearing aid dispensers. In addition, the responses measured in an anechoic chamber do not accurately reflect the real world performance that might be expected in a typical hard-walled room such as in a home or office environment where standing waves are present. It has not previously been possible to assess the performance of a directional microphone system in a real world echoic environment because it has not been possible to present appropriate front-facing and rear-facing signals simultaneously.

BRIEF SUMMARY OF THE INVENTION

[0005] Accordingly, it is an object of the invention to provide a method and apparatus for more accurately testing the directional-response of a hearing aid, even if the hearing aid has level-dependent signal processing circuitry.

[0006] In one embodiment the invention provides apparatus for testing a directional responding acoustic device, comprising:

- (a) at least first and second sound sources adapted to be placed in first and second positions respectively relative to said device,
- (b) at least one signal generator coupled to said first and second sound sources for generating a first audio signal applied to said first sound source and a second audio signal simultaneously applied to said second sound source, said first and second sound sources generating simultaneous first and second acoustical signals in response to said first and second audio signals applied thereto,

- (c) said first and second audio signals and hence said first and second acoustical signals each containing a plurality of orthogonal components, the components of said first audio signal being different from the components of said second audio signal,
- (d) and an analyzer adapted to be coupled to said device and synchronized with said generator, for analyzing the response of said device to said first and second acoustical signals.

[0007] In another embodiment the invention provides a method for testing a directional responding acoustic device, comprising:

- (a) generating at least first and second audio signals each containing a plurality of components, the components of said first audio signal being different from the components of said second audio signal and being orthogonal thereto,
- (b) applying said first and second audio signals to first and second sound sources respectively to produce first and second acoustical signals,
- (c) exposing said device simultaneously to said first and second acoustical signals to produce a received signal,
- (d) and analyzing the response of said device to said first and second acoustical signals.

[0008] Further objects and advantages of the invention will appear from the following description, taken together with the accompanying drawings.

BRIEF DESCRIPTION OF DRAWINGS

[0009] In the drawings:

Fig. 1 is a block diagrammatic view of a directional hearing aid test apparatus according to the invention;

Fig. 2 is a block diagram view of a signal generator/analyzer used in the Fig. 1 apparatus;

Fig. 3 is a plot showing the response of the Fig. 1 apparatus with a front excitation signal on;

5 Fig. 4 is a plot showing the response of the Fig. 1 apparatus with a rear excitation signal on;

Fig. 5 is a plot showing the response of the Fig. 1 apparatus with both the front and rear excitation signals on;

10 Fig. 6 is a plot showing the response of the Fig. 1 apparatus with both the front and rear excitation signals off;

Fig. 7 is a plot showing the response of the Fig. 1 apparatus to
15 the acoustical output signal from a hearing aid operating in a non-directional mode;

Fig. 8 is a plot showing the response of the Fig. 1 apparatus to the acoustical output of a directional hearing aid set to a directional mode; and
20

Fig. 9 is a diagrammatic block view showing a modified arrangement according to the invention.

DETAILED DESCRIPTION OF PREFERRED EMBODIMENTS

25 **[0010]** The preferred embodiment of the invention will be described with reference to testing a directional hearing aid. However, the method and apparatus of the invention may be used with other directional-responding acoustical devices, e.g. microphones, and sound recorders for various applications.

30 **[0011]** As shown in Fig. 1, a test space **10**, which can be either an acoustically-treated anechoic space or a non-treated echoic space, contains two spaced apart loudspeakers, namely a first speaker **12** and a second speaker **14**. The two loudspeakers are shown as being in the same plane

and facing each other, but this configuration is arbitrary and depends on the performance characteristic which is desired to be measured.

[0012] The hearing aid **16** to be tested is shown midway between the loudspeakers **12**, **14**, but the hearing aid **16** can be placed in any desired
5 orientation. Located closely adjacent the hearing aid **16** are a controlling microphone **18** (used for a purpose to be explained), a conventional ear simulator or coupler **20** which is connected to the hearing aid **16**, and a measurement microphone **22** which (via the coupler **20**) receives the
10 acoustical signal output by the hearing aid **16** (and which acoustical signal would normally be directed into a user's ear).

[0013] The loudspeakers **12**, **14** are connected to an audio signal generator **24**, to be described in more detail, and which generates audio signals to excite each loudspeaker.

[0014] The controlling microphone **18** is connected to an analyzer **26**,
15 which in turn is connected to and controls the audio signal generator **24**. The measurement microphone **22** is also connected to the analyzer **26**, for analysis of the hearing aid response.

[0015] The audio signal generator **24** is a computer controlled signal generator which is clocked by a clock diagrammatically indicated at **28**. The
20 clock **28** also provides a clock signal to the computer controlled analyzer **26**, so that the analyzer **26** is synchronized precisely to the generator **24**. In fact, the generator **24** and analyzer **26** are normally implemented as one piece of equipment, as will be disclosed.

[0016] The generator **24** generates two broadband excitation signals,
25 one for first loudspeaker **12** and the other for the second loudspeaker **14**. The first broadband signal, indicated at **30** in Fig. 1, consists of multiple sinusoids which are exact bin frequencies of a Discrete Fourier Transform ("DFT"), but signal **30** does not contain all of the bin frequencies. The second broadband signal, indicated at **32** in Fig. 1, and which is applied to the second
30 loudspeaker **14**, is composed of multiple sinusoids which are the unused bin

frequencies of the DFT of the first excitation signal **30**. Each signal **30, 32** can contain an arbitrary quantity and spacing of bin frequencies, with the important requirement that no bin frequency be common to both signals. A particularly useful configuration is to have one of the audio signals **30, 32** contain the even bin frequencies, and the other contain the odd bin frequencies, so that the bandwidths and spectra of each audio signal are very similar.

[0017] The signal **34** appearing at the output of the controlling microphone **18** is a linear combination of the two mathematically orthogonal excitation signals **30, 32**. The signal **34** is converted to the frequency domain by the signal analyzer **26** to which controlling microphone **18** is connected, using (in the signal analyzer) a DFT. Once in the frequency domain, the primary and secondary signal DFT bin components are separated by the signal analyzer **26** (a simple task), thereby extracting the received signals corresponding to the primary and secondary excitation signals **30, 32**. Since the signal analyzer **26** is connected to the signal generator **24**, independent control loops are implemented for each excitation signal **30, 32**, so that the level, phase and spectral content of each excitation signal **30, 32** can be precisely controlled.

[0018] In more detail, and as shown in Fig. 2, which shows the analyzer/generator as a single block **24/26**, the analogue signal **34** from the controlling microphone **18** is applied to an A/D converter **36** in block **24/26**. The resulting digital signal **38** is applied to a processor **40** which implements a Fast Fourier Transform or FFT (which is an efficient means by which to calculate the DFT) to convert signal **35** to two frequency domain signals **42, 44**, one containing the bin components of first excitation signal **30** and the other containing the bin components of second excitation signal **32**.

[0019] Signals **42, 44** are applied to control a signal generator processor **46** (typically the same processing hardware as FFT processor **40**) to produce two frequency domain signals **30', 32'** corresponding to excitation signals **30, 32**. Signals **30', 32'** are passed through an inverse FFT processor

48 (again part of the same processor hardware previously mentioned) to producing two time domain digital signals **30"**, **32"** corresponding to excitation signals **30**, **32** respectively. Signals **30"**, **32"** are passed through D/A converters **50**, **52** to produce the excitation signals **30**, **32** (which can be
5 appropriately amplified, by amplifiers not shown). In this way, the excitation signals **30**, **32** are controlled to have any desired characteristics. For example, the spectrum of each excitation signal **30**, **32** can be made that of "pink noise" (i.e. flat on a logarithmic scale), or the spectrum of each excitation signal can be made that of speech in a crowded room, or to have any other
10 desired shape.

[0020] Preferably, the controlling microphone **18** has a flat, non-directional response, but this is not essential since its characteristics can be compensated for as desired.

[0021] The acoustic signal (resulting from first and second excitation
15 signals **30**, **32**) which excites or drives the controlling microphone **18** is also (to a very close approximation) the same as the signal appearing at the hearing aid **16** and which is processed by the hearing aid. If the level of the excitation signals **30**, **32** is sufficiently low, the distortion at the hearing aid **16** is negligible. The hearing aid **16** outputs an acoustical signal which is
20 directed through coupler **20** to the measurement microphone **22**, which in turn outputs a received audio signal **54**. Again, the received signal **54** is essentially a linear combination of the two mathematically orthogonal excitation signals **30**, **32**. In block **24/26**, the received signal **54** is converted to a digital signal **56** by A/D converter **58**, and is then converted to a
25 frequency domain signal using FFT processor **62**. Processor **62** also separates the primary and secondary signal bin components in such frequency domain signal and provides two output signals **66**, **68**, one containing the hearing aid's response to primary excitation signal **30**, and the other containing the hearing aid's response to secondary excitation signal **32**.
30 (As before, processors **62**, **64** can be part of the same processing hardware

previously mentioned.) The output signals **66, 68** can be viewed on a monitor, or can be printed, or can otherwise be dealt with as desired.

[0022] An important feature of the invention is that each sinusoid in each of the excitation signals **30, 32** is precisely on-bin for the DFT and is therefore orthogonal to every other sinusoid. In addition, because the analyzer **26** is precisely synchronized to the generator **24** (as mentioned, they may be integrated as one hardware unit), therefore when a DFT is performed on the received signal **54** from the measurement microphone **22** (after signal **54** is converted into digital signal **56**), all the spectral components of the received signal **54** will also fall precisely on-bin, and therefore there will be no smearing of information between frequencies because they are completely orthogonal. Because the signals are orthogonal, no filtering is necessary.

[0023] If the excitation signals **30, 32** were generated in the time domain without regard to their DFT frequency bin alignment, and if the received signal were then analyzed with a DFT, the approach would work if the frequencies were sufficiently separated so that the unavoidable frequency smearing effects could be neglected. However, there would be a point at which the smearing would cause the adjacent frequencies to merge into one spectral line and become inseparable. Well-known time domain windowing techniques can reduce the frequency smearing but cannot eliminate it. There would also be unavoidable trade-offs between frequency resolution and amplitude accuracy. Ultimately, there would be severe limits to how closely frequencies can be spaced, and this would limit the ability to create a dense spectrum that can be separated by analysis. In contrast, the method and the apparatus described are less prone to these limitations and separable spectra can be created to most reasonable requirements so long as the generator and analyzer are accurately synchronized.

[0024] By way of example, an excitation signal bandwidth (for each excitation signal **30, 32**) of 200 Hz to 8 kHz can be provided. This is a bandwidth which is typically used in the measurement of hearing aids. Modest performance conventional hardware can be used which runs at a

sample rate of 32 kHz and uses a 4096 point DFT, in which case it is possible to produce approximately 1000 mathematically orthogonal sinusoids in the bandwidth of 200 Hz to 8 kHz. If half of the sinusoids (500) are allocated to the first excitation signal **30**, applied to the first speaker **12**, and 500 sinusoids
5 are allocated to the second excitation signal **32** for the second speaker **14**, then the spectrum can be divided so that odd bin frequencies are allocated to the first excitation signal **30** and even bin frequencies are allocated to the second excitation signal **32**. In that case, the spectrum for each excitation signal will have frequency components spaced apart about every 16 Hz,
10 which is sufficiently dense to meet the requirements for testing the broadband directional characteristics of current hearing aids. Some current hearing aids have processing bands as narrow as 100 Hz, so it is evident that the dense spectrum which the invention can achieve is already highly useful. Future hearing aids may require even denser spectrums, which can be achieved
15 relatively easily using the described method and apparatus.

[0025] If an application requires a different stimulus spectrum, then the sampling rate for the DFT size, or both, can be scaled to meet the requirements without serious concern about smearing or cross-talk between the excitation signals, because their components are orthogonal and the
20 orthogonality is preserved independent of the scaling, provided that the generator **24** and analyzer **26** are synchronized.

[0026] It will be realized that the bin allocations can be changed from the odd/even arrangement described in the example, depending on the desired characteristics to be measured. For example, if a less dense
25 spectrum is required for the second excitation signal **32**, then (by way of example only), two-thirds of the bin frequencies can be allocated to the first excitation signal **30** and one of each three can be allocated to the second excitation signal **32**.

[0027] Reference is next made to Figs. 3 to 8, which show
30 experimental results generated from the system previously described. To produce the experimental results, the first and second excitation signals **30**,

32 were applied simultaneously (to speakers **12** and **14** respectively), and each consisted of approximately 500 sinusoids. Each excitation signal was controlled to have an overall level of 60 dBSPL over a bandwidth of 200 Hz to 8000 Hz as measured by the controlling microphone **18**. In Figs. 3 to 8, the
5 X-axis units are Hz, and the Y-axis are dBSPL.

[0028] The responses of Figs. 3 to 6 were measured at the measurement microphone **22**, without a hearing aid present. All measurements for Figs. 3 to 6 were performed without a coupler attached to the measurement microphone **22**.

10 [0029] The responses in Figs. 3 to 8 are shown in $\frac{1}{12}$ th octave bands. Since there are a total of 65 such bands in the bandwidth between 200 and 8000 Hz, therefore the response curves are each made up of 65 points.

[0030] In Fig. 3, the audio signal **30** exciting the front speaker **12** was on, while the audio signal **32** exciting the rear speaker **14** was off. It will be
15 seen that the response **70** resulting from signal **32** accurately measures the 60 dBSPL stimulus, while the response **72** from the rear speaker **14** is shown at **72** and measures the noise floor of the device. There was no interaction between the front and rear signals **30**, **32**.

[0031] In Fig. 4, the opposite situation prevailed. Rear signal **32** which
20 fed rear speaker **14** was on, while front signal **30** feeding front speaker **12** was off. Curve **74** accurately measures the 60 dBSPL stimulus from the rear speaker **14**, while curve **76** resulting from the lack of any signal from the front speaker **12** shows that the device was measuring its noise floor in respect of any signal from the front speaker. There was no interaction between the rear
25 and front signals.

[0032] For Fig. 5, both the front and rear signals **30**, **32** were on and controlled for the front and rear speakers **12**, **14** to output an overall level of 60 dBSPL. The two measured responses, commonly indicated at **78**, are essentially overlays (as expected), and it will be seen that they do not interact
30 with each other.

[0033] For Fig. 6, both the front and rear signals **30, 32** were off and the device measured its noise floor as shown by the front and rear response curves **80, 82**.

[0034] For Fig. 7, the measurement microphone **22** was connected to
5 an ANSI HA-2 hearing aid coupler **20**. The HA-2 coupler simulates the volume of an average human ear canal. The coupler **20** was connected to a Phonak P2AZ directional hearing aid and the hearing aid was set to its omni-directional (i.e. non-directional) mode. The front and rear response curves are shown at **84, 86**, and as expected, they are essentially overlays, i.e. no
10 directional response was seen.

[0035] Fig. 8 displays the response of the same Phonak P2AZ directional hearing aid when set to its directional mode. The difference in responses to the front and rear signals can clearly be seen in curves **88, 90**.

[0036] While only two speakers **12, 14** have been shown, driven by two
15 excitation signals **30, 32**, the number of speakers can be increased, and the number of excitation signals can also be increased, for example to provide a different excitation signal for each speaker, or to drive two or more speakers with the same excitation signal. As before, the components of each excitation signal will always be orthogonal to each other.

[0037] For example, four speakers can be used, or alternatively the
20 directionality characteristics can be measured at quadrature position points on a sphere, such measurement being in real time. An example is shown in Fig. 9, where the hearing aid to be tested, a controlling microphone, and the measurement microphone and coupler connecting it to the hearing aid, are all
25 indicated at block **92**. Four speakers **94, 96, 98, 100** are provided, one in front of the hearing aid, one behind it, and one at each side. Each speaker is preferably excited with an excitation signal having bin frequencies different from the bin frequencies of each of the other excitation signals exciting the other speakers. Because the excitation signals are therefore all orthogonal to
30 each other, the response to each excitation signal can easily be separated from the other responses, without filtering.

[0038] It will be understood that the term "sinusoid" as used in this description means a signal having the shape of a sine wave, but having any desired amplitude and phase. For example, "sinusoid" includes a cosine wave.

5 **[0039]** In addition, while it has been assumed that the front and rear signals **30, 32** in the example given both contain bin frequencies from the same DFT, in fact different DFTs can be used, so long as one is an integer multiple or sub-multiple of the other. For example, one can be four times as dense as the other, in which case one of every four bins would coincide. For
10 the coincident bins, only one excitation signal would have a bin frequency from that bin, so that in no cases would the front and rear excitation signals contain any of the same bin frequencies. Since the bin frequencies of the front and rear excitation signals would remain different, the front and rear excitation signals would be orthogonal to each other as before. However, one
15 excitation signal would be much denser than the other.

[0040] While sinusoidal wave forms are preferred for the components of the excitation signals, since sinusoids are easy to generate and are orthogonal, other orthogonal signals can be used. For example, Walsh Transforms, which provide square waves, can be used, provided that
20 appropriate square waves are selected so that the square waves of one excitation signal are orthogonal to those of the other.

[0041] Alternatively, the excitation signals can employ wavelets, or any other orthogonal components.

[0042] It will be seen that using a preferred embodiment of the
25 invention, the response of a hearing aid can be tested in relatively "real world" conditions, e.g. non-anechoic environments, even where the hearing aid has non-linear and level dependent signal processing circuitry. Since in the preferred embodiment of the invention the primary and secondary excitation signals are presented simultaneously, the level dependent circuitry in the
30 hearing aid **16** is properly excited for assessing the hearing aid response characteristics.

[0043] In addition, the hearing aid response can be displayed in real time, so that changes to the directional characteristics can be quickly evaluated.

[0044] While normally, it is the acoustic output signal from a hearing aid
5 that will be evaluated, in some cases (e.g. where the hearing aid is under development), its electrical output signal will be available and can be evaluated using the apparatus and method described.

[0045] While preferred aspects of the invention have been described, it
10 will be understood that various changes can be made within the scope of the invention.